

STD100NH02L

N-CHANNEL 24V - 0.0042 Ω - 60A DPAK/IPAK STripFETTM III POWER MOSFET

TYPE	V _{DSS}	R _{DS(on)}	I _D
STD100NH02L	24 V	< 0.0048 Ω	60 A(2)

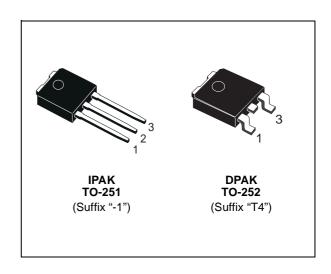
- TYPICAL $R_{DS}(on) = 0.0042 \Omega @ 10 V$
- TYPICAL $R_{DS}(on) = 0.005 \Omega @ 5 V$
- R_{DS(ON)} * Qg INDUSTRY's BENCHMARK
- CONDUCTION LOSSES REDUCED
- SWITCHING LOSSES REDUCED
- LOW THRESHOLD DEVICE
- THROUGH-HOLE IPAK (TO-251) POWER PACKAGE IN TUBE (SUFFIX "-1")
- SURFACE-MOUNTING DPAK (TO-252)
 POWER PACKAGE IN TAPE & REEL (SUFFIX "T4")

DESCRIPTION

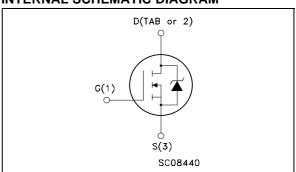
The STD100NH02L utilizes the latest advanced design rules of ST's proprietary STripFET™ technology. This is suitable fot the most demanding DC-DC converter application where high efficiency is to be achieved.

APPLICATIONS

 SPECIFICALLY DESIGNED AND OPTIMISED FOR HIGH EFFICIENCY DC/DC CONVERTES



INTERNAL SCHEMATIC DIAGRAM



Ordering Information

SALES TYPE	MARKING	PACKAGE	PACKAGING
STD100NH02LT4	D100NH02L	TO-252	TAPE & REEL
STD100NH02L-1	D100NH02L	TO-251	TUBE

ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Value	Unit
V _{spike(1)}	Drain-source Voltage Rating	30	V
V _{DS}	Drain-source Voltage (V _{GS} = 0)	24	V
V_{DGR}	Drain-gate Voltage ($R_{GS} = 20 \text{ k}\Omega$)	24	V
V _{GS}	Gate- source Voltage	± 20	V
I _D (2)	Drain Current (continuous) at T _C = 25°C	60	Α
I _D (2)	Drain Current (continuous) at T _C = 100°C	60	Α
I _{DM} (3)	Drain Current (pulsed)	240	Α
P _{tot}	Total Dissipation at T _C = 25°C	100	W
	Derating Factor	0.67	W/°C
E _{AS} (4)	Single Pulse Avalanche Energy	800	mJ
T _{stg}	Storage Temperature	-55 to 175	°C
Tj	Max. Operating Junction Temperature	-55 to 175	

September 2003 1/12

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THERMAL DATA

1.5 100 275	°C/W °C	

ELECTRICAL CHARACTERISTICS ($T_{CASE} = 25~^{\circ}C$ UNLESS OTHERWISE SPECIFIED) OFF

Symbol	Parameter	Test Conditions Min.		Тур.	Max.	Unit
V _{(BR)DSS}	Drain-source Breakdown Voltage	$I_D = 25 \text{ mA}, V_{GS} = 0$	24			V
I _{DSS}	Zero Gate Voltage Drain Current (V _{GS} = 0)	V _{DS} = 20 V V _{DS} = 20 V T _C = 125°C			1 10	μA μA
I _{GSS}	Gate-body Leakage Current (V _{DS} = 0)	V _{GS} = ± 20V			±100	nA

ON (5)

Symbol	Parameter	Test Conditions		Min.	Тур.	Max.	Unit
V _{GS(th)}	Gate Threshold Voltage	$V_{DS} = V_{GS}$	I _D = 250 μA	1	1.8		V
R _{DS(on)}	Static Drain-source On Resistance	V _{GS} = 10 V V _{GS} = 5 V	I _D = 30 A I _D = 15 A		0.0042 0.005	0.0048 0.009	Ω Ω

DYNAMIC

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
gfs (5)	Forward Transconductance	V _{DS} = 10 V I _D = 30 A		50		S
C _{iss} C _{oss} C _{rss}	Input Capacitance Output Capacitance Reverse Transfer Capacitance	$V_{DS} = 15V f = 1 MHz V_{GS} = 0$		3940 1020 110		pF pF pF
R _G	Gate Input Resistance	f = 1 MHz Gate DC Bias = 0 Test Signal Level = 20 mV Open Drain		1.1		Ω

ELECTRICAL CHARACTERISTICS (continued)

SWITCHING ON

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
t _{d(on)} t _r	Turn-on Delay Time Rise Time	$\begin{aligned} V_{DD} &= 10 \text{ V} & I_D &= 30 \text{ A} \\ R_G &= 4.7 \Omega & V_{GS} &= 10 \text{ V} \\ \text{(Resistive Load, Figure 3)} \end{aligned}$		15 200		ns ns
Q _g Q _{gs} Q _{gd}	Total Gate Charge Gate-Source Charge Gate-Drain Charge	V _{DD} = 10 V I _D = 60 A V _{GS} = 10 V		62 12 8	84	nC nC nC
Q _{oss} (6)	Output Charge	V _{DS} = 16 V V _{GS} = 0 V		24		nC
Q _{gls} (7)	Third-quadrant Gate Charge	V _{DS} < 0 V V _{GS} = 10 V		56.5		nC

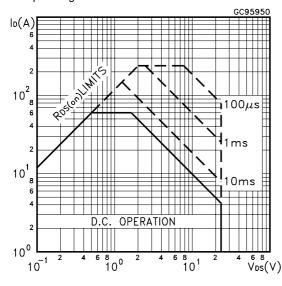
SWITCHING OFF

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
t _{d(off)}	Turn-off Delay Time Fall Time	$ \begin{array}{llllllllllllllllllllllllllllllllllll$		60 35	47	ns ns

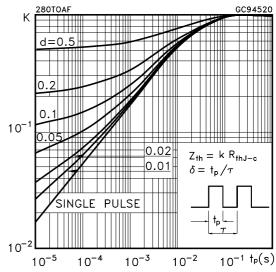
SOURCE DRAIN DIODE

Symbol	Parameter	Test Conditions	Min.	Тур.	Max.	Unit
I _{SD} I _{SDM}	Source-drain Current Source-drain Current (pulsed)				60 240	A A
V _{SD} (5)	Forward On Voltage	I _{SD} = 30 A V _{GS} = 0			1.3	V
t _{rr} Q _{rr} I _{RRM}	Reverse Recovery Time Reverse Recovery Charge Reverse Recovery Current	$I_{SD} = 60 \text{ A}$ di/dt = $100 \text{A}/\mu \text{s}$ $V_{DD} = 15 \text{ V}$ $T_j = 150 ^{\circ} \text{C}$ (see test circuit, Figure 5)		47 58 2.5		ns nC A

Safe Operating Area



Thermal Impedance

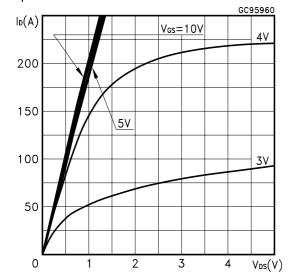


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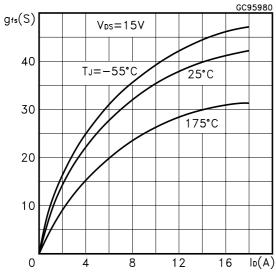
⁽¹⁾ Garanted when external Rg=4.7 Ω and $t_f < t_{fmax}$. (2) Value limited by wire bonding (3) Pulse width limited by safe operating area. (4) Starting $T_j = 25$ °C, $I_D = 30$ Å, $V_{DD} = 15$ V

⁽⁵⁾ Pulsed: Pulse duration = 300 μ s, duty cycle 1.5 %. (6) $Q_{OSS} = C_{OSS}^* \Delta \ V_{in}$, $C_{OSS} = C_{gd} + C_{ds}$. See Appendix A (7) Gate charge for synchronous operation

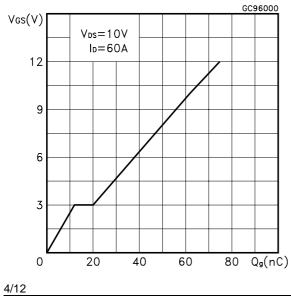
Output Characteristics



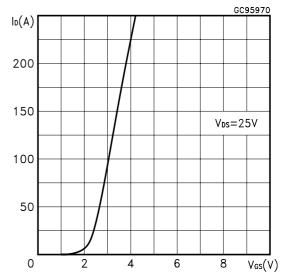
Transconductance



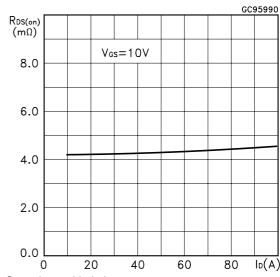
Gate Charge vs Gate-source Voltage



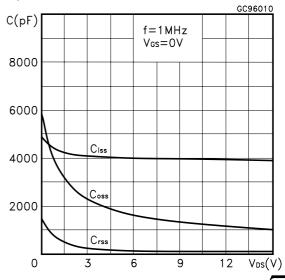
Transfer Characteristics



Static Drain-source On Resistance

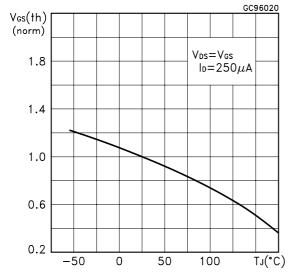


Capacitance Variations

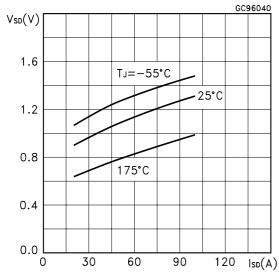


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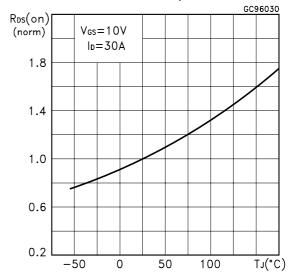
Normalized Gate Threshold Voltage vs Temperature



Source-drain Diode Forward Characteristics



Normalized on Resistance vs Temperature



Normalized Breakdown Voltage vs Temperature

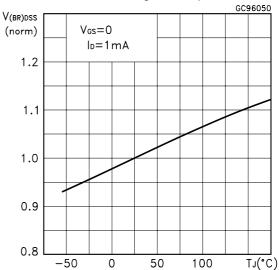


Fig. 1: Unclamped Inductive Load Test Circuit

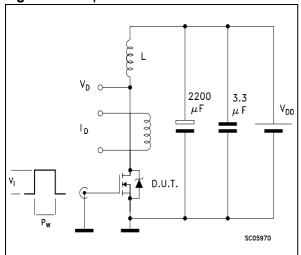


Fig. 3: Switching Times Test Circuits For Resistive Load

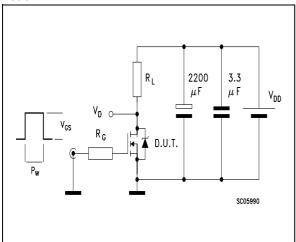


Fig. 5: Test Circuit For Inductive Load Switching And Diode Recovery Times

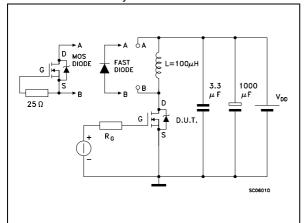


Fig. 2: Unclamped Inductive Waveform

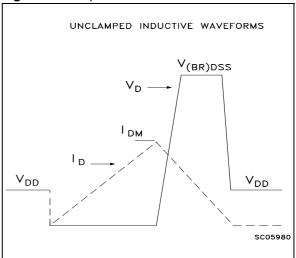
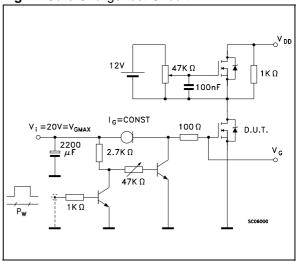
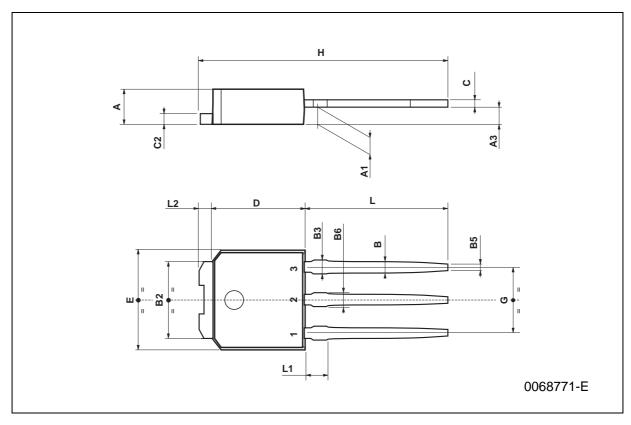


Fig. 4: Gate Charge test Circuit



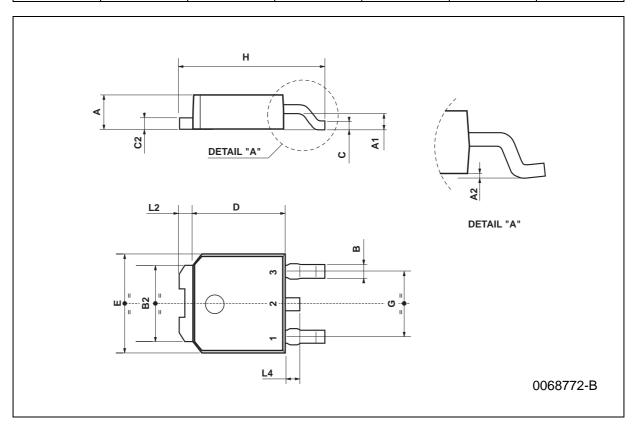
TO-251 (IPAK) MECHANICAL DATA

DIM.		mm			inch	
DIIVI.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
А	2.2		2.4	0.086		0.094
A1	0.9		1.1	0.035		0.043
А3	0.7		1.3	0.027		0.051
В	0.64		0.9	0.025		0.031
B2	5.2		5.4	0.204		0.212
В3			0.85			0.033
B5		0.3			0.012	
B6			0.95			0.037
С	0.45		0.6	0.017		0.023
C2	0.48		0.6	0.019		0.023
D	6		6.2	0.236		0.244
Е	6.4		6.6	0.252		0.260
G	4.4		4.6	0.173		0.181
Н	15.9		16.3	0.626		0.641
L	9		9.4	0.354		0.370
L1	0.8		1.2	0.031		0.047
L2		0.8	1		0.031	0.039



TO-252 (DPAK) MECHANICAL DATA

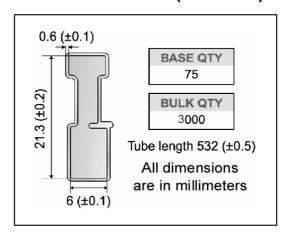
DIM.		mm		inch		
Diwi.	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
А	2.2		2.4	0.086		0.094
A1	0.9		1.1	0.035		0.043
A2	0.03		0.23	0.001		0.009
В	0.64		0.9	0.025		0.035
B2	5.2		5.4	0.204		0.212
С	0.45		0.6	0.017		0.023
C2	0.48		0.6	0.019		0.023
D	6		6.2	0.236		0.244
Е	6.4		6.6	0.252		0.260
G	4.4		4.6	0.173		0.181
Н	9.35		10.1	0.368		0.397
L2		0.8			0.031	
L4	0.6		1	0.023		0.039



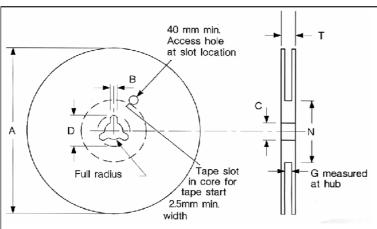
DPAK FOOTPRINT

6.7 1.8 3.0 1.6 2.3 1.6 All dimensions are in millimeters

TUBE SHIPMENT (no suffix)*



TAPE AND REEL SHIPMENT (suffix "T4")*



REEL MECHANICAL DATA

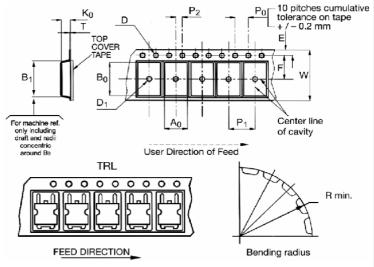
DIM.	mm		inch	
	MIN.	MAX.	MIN.	MAX.
Α		330		12.992
В	1.5		0.059	
С	12.8	13.2	0.504	0.520
D	20.2		0.795	
G	16.4	18.4	0.645	0.724
N	50		1.968	
Т		22.4		0.881

TAPE MECHANICAL DATA

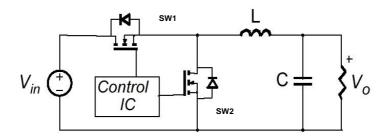
	mm inch				
DIM.					
	MIN.	MAX.	MIN.	MAX.	
A0	6.8	7	0.267	0.275	
B0	10.4	10.6	0.409	0.417	
B1		12.1		0.476	
D	1.5	1.6	0.059	0.063	
D1	1.5		0.059		
E	1.65	1.85	0.065	0.073	
F	7.4	7.6	0.291	0.299	
K0	2.55	2.75	0.100	0.108	
P0	3.9	4.1	0.153	0.161	
P1	7.9	8.1	0.311	0.319	
P2	1.9	2.1	0.075	0.082	
R	40		1.574		
W	15.7	16.3	0.618	0.641	

 BASE QTY
 BULK QTY

 1000
 1000



APPENDIX A Buck Converter: Power Losses Estimation



The power losses associated with the FETs in a Synchronous Buck converter can be estimated using the equations shown in the table below. The formulas give a good approximation, for the sake of performance comparison, of how different pairs of devices affect the converter efficiency. However a very important parameter, the working temperature, is not considered. The real device behavior is really dependent on how the heat generated inside the devices is converted to allow for a safer working junction temperature.

The low side (SW2) device requires:

- ullet Very low $R_{DS(on)}$ to reduce conduction losses
- $\bullet \qquad Small \ Q_{gls} \ to \ reduce \ the \ gate \ charge \ losses$
- Small Coss to reduce losses due to output capacitance
- Small Q_{rr} to reduce losses on SW₁ during its turn-on
- The C_{gd}/C_{gs} ratio lower than V_{th}/V_{gg} ratio especially with low drain to source voltage to avoid the cross conduction phenomenon;

The high side (SW1) device requires:

- ullet Small R_g and L_s to allow higher gate current peak and to limit the voltage feedback on the gate
- Small Qg to have a faster commutation and to reduce gate charge losses
- $\bullet \qquad \text{Low } R_{DS(on)} \text{ to reduce the conduction losses}.$

		High Side Switch (SW1)	Low Side Switch (SW2)
Pconduct	ion	$R_{DS(on)SW1} * I_L^2 * d$	$R_{DS(on)SW2} * I_L^2 * (1-d)$
Pswitchin	g	$V_{\text{in}} * (Q_{\text{gsth(SW1)}} + Q_{\text{gd(SW1)}}) * f * \frac{I_L}{I_g}$	Zero Voltage Switching
P_{diode}	Recovery	Not Applicable	¹ V _{in} *Q _{rr(SW2)} * f
	Conduction	Not Applicable	$V_{\text{f(SW2)}} * I_{\text{L}} * t_{\text{deadtime}} * f$
Pgate(QG)	$Q_{g(SW1)} * V_{gg} * f$	$Q_{gls(SW2)}*V_{gg}*f$
P _{Qoss}		$\frac{V_{in} *Q_{oss(SW1)} *f}{2}$	$\frac{V_{\text{in}} * Q_{\text{oss(SW2)}} * f}{2}$

Parameter	Meaning	
d	Duty-cycle	
Qgsth	Post threshold gate charge	
$Q_{ m gls}$	Third quadrant gate charge	
Pconduction	On state losses	
Pswitching	On-off transition losses	
Pdiode	Conduction and reverse recovery diode losses	
Pgate	Gate drive losses	
P _{Qoss}	Output capacitance losses	

¹ Dissipated by SW1 during turn-on

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